# Landing Control Based on Energy Prediction for a Quadcopter Under External Disturbances

Cuong V. Nguyen <sup>1\*</sup>

<sup>1</sup> Thai Nguyen University of Information and Communication Technology (ICTU), Viet Nam Email:  $\frac{1}{1}$  nvcuong@ictu.edu.vn \*Corresponding Author

*Abstract***—Unmanned aerial vehicles (UAVs) have recently become one of the most popular research topics. The high diversity in its uses has attracted research attention regarding structure or control capabilities. However, if the energy consumed in each mission cannot be predicted, the available flight time will pose many risks to the UAV and data security. This paper proposes a control algorithm based on predicting the remaining flight time to determine a safe landing station. Suppose the UAV cannot reach the desired destination station. In that case, it will find the nearest landing station to recharge its energy until the SoC (State of Charge) > 90%, then the UAV will continue to perform the mission until the UAV reaches the destination station. In addition, the paper uses a marker-based landing method to improve landing accuracy. The sliding mode controller (SMC) is designed to consider external disturbance factors and consider a solution to reduce chattering.**

# *Keywords—SMC; UAV; Control; Landing; Energy Prediction.*

# I. INTRODUCTION

In recent years, the research of unmanned aerial vehicles (UAVs) for civil or military applications has promoted the need to operate these systems with higher requirements. The UAVs have shown versatility as well as efficiency in various fields such as search and rescue (SAR) [1][2], meteorological research [3]-[6], infrastructure testing [7]-[10], homeland security and traffic monitoring [11]-[15], and precision agriculture [16]-[20]. With a large number of quadcopters in use, it is undeniable that quadcopters are capable of because of their convenience and flexibility in narrow operating ranges. They can fly in low areas, hover, and provide detailed information about that area through the control station [21][22].

Unmanned aerial vehicles (UAVs) in the form of quadcopters are a popular research topic. The ability to perform various aerial maneuvers opens up many opportunities for creating applications such as surveillance systems, delivery, video recording, or other military purposes. However, the success rate of such applications depends on the flight time in relation to energy consumption and battery capacity [23]-[25].

UAVs (Unmanned Aerial Vehicles), also known as drones, can encounter many serious risks when operating and encountering a situation where the battery runs out [26]. Here are some of the main risks:

**Free fall:** If the UAV's battery suddenly runs out, it can completely lose control and fall freely, causing damage to the UAV, property, or people below.

**Loss of control:** A weak battery can cause the UAV to lose its ability to maintain stability and control, leading to uncontrolled movement and the risk of collision with other objects.

**Data loss:** If the UAV is performing a recording or data collection mission, a sudden battery drain can cause the loss of data that has not been stored yet.

**Economic Loss:** A crashed or damaged UAV can result in high repair or replacement costs, affecting the economic viability of the missions the UAV is performing.

Therefore, proper battery testing and management, as well as using the PIN SoC monitoring and prediction system to issue emergency landing control commands, are essential to minimize risks.

There are many methods to control a quadcopter to follow a given trajectory such as the PID method [27]-[30]. In the article [31]-[35], the author proposes a Backstepping controller to control the position and attitude of the quadcopter. The sliding control method is one of the popular methods used to control a quadcopter as presented in the article [36]-[41]. In the article [42]-[45], the author uses neural network method combined with other methods to build a quadcopter controller. However, this method requires a huge computing capacity. Linear controllers such as PID and LQR are not good for controlling a strongly nonlinear object like a quadcopter. In this article we build an SMC controller to control a quadcopter to follow a trajectory and land based on energy prediction.

The landing phase is the last and most important phase of a UAV's navigation process [46]. Therefore, landing methods for UAVs are attracting considerable attention from researchers worldwide. Previous studies have proposed global positioning systems (GPS) and inertial navigation sensors (INS) as the main positioning systems [47]-[49]. However, data from the GPS often has large deviations. Therefore, the landing location is often 1 to 3 m from the desired location [50]. Based on the above issues, this work aims to develop a new marker-based landing system that uses a sliding mode control to control the UAV's trajectory and land in a specific location with great precision. The landing position error of the UAV has been improved thanks to the use of ArUco markers [51]-[58].



Research and development of methods to determine optimal landing points based on the remaining battery SoC for quadcopters is a topic that is attracting the attention of many researchers around the world. In [59], the authors studied an energy-saving landing planning method for UAVs based on SoC and wind speed predictions, determining the optimal quadcopter landing trajectory under changing weather conditions. In [60], the authors studied how to select the optimal landing point for quadcopters based on the remaining SoC of the battery and ground obstacles. In [61][62], the authors focused on optimal trajectory planning and energy saving for quadcopters during landing. In [63][64], an overview of the methods and challenges in energy-saving flight for quadcopters, including SoC-based landing planning.

The contribution of this paper is given by the following points: i) Building an SMC controller based on Lyapunov stability theory considers external disturbances. ii) Building a flight and landing algorithm based on real-time energy consumption prediction of the quadcopter, thereby creating a basis for building a control scheme in different modes. That has helped the quadcopter avoid risks such as collision with stations, loss, etc., and supported the quadcopter in operating more effectively. iii)The UAV's landing process is based on the marker maker instead of using the signal sent back from the GPS system, so the accuracy is significantly improved during the experiment [65]-[67].

The rest of this paper is organized as follows. The kinematic model of the Quadcopter is presented in Section 2. The SMC for position control and attitude control is described in Section 3. Landing control based on energy prediction shown in Section 4. Finally, conclusions are provided in Section 5.

#### II. DYNAMIC MODELLING FOR A QUADCOPTER

The structural model of the quadrotor and the coordinate systems used in building the dynamics model of the quadrotor are shown in Fig. 1.



Fig. 1. Structural model of Quadrotor

Using the Newton - Euler method for the model [68], we get the quadrotor's equations of motion as (1).

$$
\begin{cases} m\ddot{\Gamma}^E = F_{th} - F_d - F_g - F_n \\ I\ddot{\Theta}^E = M - M_{gp} - M_{gb} - M_n \end{cases} \tag{1}
$$

where  $F_{th} = R_{\Theta}(\phi, \theta, \psi) [0, 0] \sum_{i=1}^{4} F_i ]^T$  is the total thrust of the four propellers;  $F_d = diag(k_1, k_2, k_3) \dot{\Gamma}^T$  is the air resistance against the quadcopter's motion;  $F_g = [0, 0, mg]^T$ is the force of gravity;  $F_n$  is the total resistance of the noise types not included in the above components.  $M =$  $[M_{\phi}, M_{\theta}, M_{\psi}]^{T}$  is the sum of the moments of the roll, pitch, and yaw angles;  $M_{gp}$  and  $M_{gb}$  are the gyroscope moments;  $M_n$  is the resisting moment component of the disturbance, not included in the above components;  $k_1$ ,  $k_2$ , and  $k_3$  are drag coefficients.

Substituting the position vector and force components into expression (1), we get the quadcopter's translational dynamics equation as (2).

$$
\begin{cases}\n\ddot{x} = -\frac{k_x}{m}\dot{x} + (cos\phi sin\theta cos\psi + sin\phi sin\psi)\frac{v_1}{m} - \frac{F_{nx}}{m} \\
\dot{y} = -\frac{k_y}{m}\dot{y} + (cos\phi sin\theta sin\psi + sin\phi cos\psi)\frac{v_1}{m} - \frac{F_{ny}}{m} \\
\ddot{z} = -\frac{k_z}{m}\dot{z} - g + (cos\phi cos\theta)\frac{v_1}{m} - \frac{F_{nz}}{m}\n\end{cases}
$$
\n(2)

where  $F_{nx}$ ,  $F_{ny}$ ,  $F_{nz}$  are the repulsive forces of the disturbance in the  $x$ ,  $y$ , and  $z$  directions. Substituting the moment components into Equation (2) of (1), we get the kinematics equation for the rotation of the quadcopter as (3).

$$
\begin{cases}\n\ddot{\phi} = \dot{\theta}\dot{\psi}\frac{l_{y}-l_{z}}{l_{x}} - \frac{l_{p}}{l_{x}}\Omega_{\Sigma}\dot{\theta} + lb\frac{v_{2}}{l_{x}} - \frac{M_{n\phi}}{l_{x}} \\
\ddot{\theta} = \dot{\phi}\dot{\psi}\frac{l_{z}-l_{x}}{l_{y}} + \frac{l_{p}}{l_{y}}\Omega_{\Sigma}\dot{\phi} + lb\frac{v_{3}}{l_{y}} - \frac{M_{n\theta}}{l_{y}} \\
\ddot{\psi} = \dot{\phi}\dot{\theta}\frac{l_{x}-l_{y}}{l_{z}} + d\frac{v_{4}}{l_{z}} - \frac{M_{n\psi}}{l_{z}}\n\end{cases}
$$
\n(3)

where  $\Omega_{\Sigma} = \Omega_1 - \Omega_2 + \Omega_3 - \Omega_4$  is the total speed of the propellers;  $M_{n\phi}$ ,  $M_{n\theta}$ ,  $M_{n\psi}$  are the resisting moment components of the disturbance acting on the roll, pitch, and yaw angles. The complete system of kinetic equations of the Quadcopter in the presence of external disturbances is expressed as (4).

$$
\begin{cases}\n\ddot{x} = -\frac{k_x}{m}\dot{x} + (\cos\phi\sin\theta\cos\psi + \sin\phi\sin\psi)\frac{U_1}{m} - \frac{F_{nx}}{m} \\
\dot{y} = -\frac{k_y}{m}\dot{y} + (\cos\phi\sin\theta\sin\psi + \sin\phi\cos\psi)\frac{U_1}{m} - \frac{F_{ny}}{m} \\
\dot{z} = -\frac{k_z}{m}\dot{z} - g + (\cos\phi\cos\theta)\frac{U_1}{m} - \frac{F_{nz}}{m} \\
\ddot{\phi} = \dot{\theta}\dot{\psi}\frac{I_y - I_z}{I_x} - \frac{I_p}{I_x}\Omega_{\Sigma}\dot{\theta} + lb\frac{U_2}{I_x} - \frac{M_{n\theta}}{I_x} \\
\ddot{\theta} = \dot{\phi}\dot{\psi}\frac{I_z - I_x}{I_y} + \frac{I_p}{I_y}\Omega_{\Sigma}\dot{\phi} + lb\frac{U_3}{I_y} - \frac{M_{n\theta}}{I_y} \\
\ddot{\psi} = \dot{\phi}\dot{\theta}\frac{I_x - I_y}{I_z} + d\frac{U_4}{I_z} - \frac{M_{n\psi}}{I_z}\n\end{cases}
$$
\n(4)

We let  $X^T = (x, y, z, \dot{x}, \dot{y}, \dot{z}, \phi, \theta, \psi, \dot{\phi}, \dot{\theta}, \dot{\psi})^T$  be the vector of state variables. The input is the propeller speed given by the equation (5).

 $\overline{\mathcal{L}}$  $\mathbf{I}$  $\mathbf{I}$  $\mathbf{I}$  $\overline{1}$ 

$$
U_1 = b(\Omega_1^2 + \Omega_2^2 + \Omega_3^2 + \Omega_4^2)
$$
  
\n
$$
U_2 = lb(-\Omega_2^2 + \Omega_4^2)
$$
  
\n
$$
U_3 = lb(-\Omega_1^2 + \Omega_3^2)
$$
  
\n
$$
U_4 = d(\Omega_1^2 - \Omega_2^2 + \Omega_3^2 - \Omega_4^2)
$$
  
\n
$$
\Omega_{\Sigma} = \Omega_1 - \Omega_2 + \Omega_3 - \Omega_4
$$
\n(5)

The equation of state describing the kinematics of the quadcopter has the following form (6)

$$
\begin{cases}\n\dot{x}_1 = x_2 \\
\dot{x}_2 = -\frac{k_x}{m}\dot{x} + (cos\phi sin\theta cos\psi + sin\phi sin\psi)\frac{U_1}{m} - \frac{F_{nx}}{m} \\
\dot{x}_3 = x_4 \\
\dot{x}_4 = -\frac{k_y}{m}\dot{y} + (cos\phi sin\theta sin\psi + sin\phi cos\psi)\frac{U_1}{m} - \frac{F_{ny}}{m} \\
\dot{x}_5 = x_6 \\
\dot{x}_6 = -\frac{k_z}{m}\dot{z} - g + (cos\phi cos\theta)\frac{U_1}{m} - \frac{F_{nz}}{m} \\
\dot{x}_7 = x_8 \\
\dot{x}_8 = \dot{\theta}\dot{\psi}\frac{I_y - I_z}{I_x} - \frac{I_p}{I_x}\Omega_{\Sigma}\dot{\theta} + lb\frac{U_2}{I_x} - \frac{M_{n\theta}}{I_x} \\
\dot{x}_9 = x_{10} \\
\dot{x}_{10} = \dot{\phi}\dot{\psi}\frac{I_z - I_x}{I_y} + \frac{I_p}{I_y}\Omega_{\Sigma}\dot{\phi} + lb\frac{U_3}{I_y} - \frac{M_{n\theta}}{I_y} \\
\dot{x}_{11} = x_{12} \\
\dot{x}_{12} = \dot{\phi}\dot{\theta}\frac{I_x - I_y}{I_z} + \frac{dU_4}{I_z} - \frac{M_{n\psi}}{I_z}\n\end{cases}
$$
\n(6)

From the system of equations of state of Quadcopter (6) we set:

 $h_1 = -\frac{k_x}{m}$  $\frac{k_x}{m}$ ;  $h_2 = -\frac{F_{nx}}{m}$  $\frac{x_{nx}}{m}$ ;  $h_3 = -\frac{k_y}{m}$  $\frac{k_y}{m}$ ;  $h_4 = -\frac{F_{ny}}{m}$  $\frac{k_{ny}}{m}$ ;  $h_5 = -\frac{k_{2}}{m}$  $\frac{n_z}{m}$ ;  $h_6 =$  $-\frac{F_{nz}}{m}$  $\frac{F_{nz}}{m}$ ;  $h_7 = \frac{I_y - I_z}{I_x}$  $\frac{1-I_z}{I_x}; h_8 = -\frac{J_r}{I_x}$  $\frac{J_r}{I_x} \Omega_{\Sigma}$ ;  $h_9 = lb \frac{1}{I_x}$ ;  $h_{10} = -\frac{M_{n0}}{I_x}$  $\frac{a_{n\emptyset}}{I_x}$ ;  $h_{11} =$  $I_z - I_x$  $\frac{I_{x}}{I_{y}}$ ;  $h_{12} = \frac{J_{p}}{I_{y}}$  $\frac{J_p}{I_y} \Omega_{\Sigma}$ ;  $h_{13} = \frac{lb}{I_y}$  $\frac{lb}{l_y}$ ;  $h_{14} = -\frac{M_{n\theta}}{l_y}$  $\frac{d_{n\theta}}{I_y}$ ;  $h_{15} = \frac{I_x - I_y}{I_z}$  $\frac{q_2}{I_z}$ ;  $h_{16} =$ d  $\frac{d}{l_z}$ ;  $h_{17} = -\frac{M_{n\psi}}{l_z}$  $\frac{dw}{I_z}$ ;  $\Omega_{\Sigma} = \omega_1 - \omega_2 + \omega_3 - \omega_4$ ;  $U_x = \frac{U_1}{m}$  $\frac{U_1}{m}$ (cos $\emptyset$ sin $\theta$ cos $\psi$  + sin $\emptyset$ sin $\psi$ );  $U_z = \frac{U_1}{m}$  $\frac{U_1}{m}(cos\phi cos\theta) - g;$  $U_y = \frac{U_1}{m}$  $\frac{U_1}{m}(cos\phi sin\theta sin\psi + sin\phi cos\psi);$ 

Then equation (6) is rewritten as (7).

$$
\begin{cases}\n\dot{x}_1 = x_2 \\
\dot{x}_2 = h_1 x_2 + h_2 + U_x \\
\dot{x}_3 = x_4 \\
\dot{x}_4 = h_3 x_4 + h_4 + U_y \\
\dot{x}_5 = x_6 \\
\dot{x}_6 = h_5 x_6 + h_6 + U_z \\
\dot{x}_7 = x_8 \\
\dot{x}_8 = h_7 x_{10} x_{12} + h_8 x_{10} + h_{10} + h_9 U_2 \\
\dot{x}_9 = x_{10} \\
\dot{x}_{10} = h_{11} x_8 x_{12} + h_{12} x_8 + h_{14} + h_{13} U_3 \\
\dot{x}_{11} = x_{10} \\
\dot{x}_{12} = h_{15} x_8 x_{10} + h_{17} + h_{16} U_4\n\end{cases} (7)
$$

# III. CONTROLLER DESIGN FOR QUADCOPTER

In this section, we have built a control scheme to control the position and attitude of the quadcopter using SMC. Position and attitude controllers were constructed, and the stability of the proposed controller was also proven.



Fig. 2. Block diagram of the quadcopter channel control system

# *A. Sliding Mode Controller Design for Position Attitude Control Quadcopter.*

We set the tracking error:

$$
\begin{bmatrix} e_1 \\ e_3 \\ e_5 \end{bmatrix} = \begin{bmatrix} x_1 - x_{1d} \\ x_3 - x_{3d} \\ x_5 - x_{5d} \end{bmatrix} \Longrightarrow \begin{bmatrix} \dot{e}_1 \\ \dot{e}_3 \\ \dot{e}_5 \end{bmatrix} = \begin{bmatrix} \dot{x}_1 - \dot{x}_{1d} \\ \dot{x}_3 - \dot{x}_{3d} \\ \dot{x}_5 - \dot{x}_{5d} \end{bmatrix} = \begin{bmatrix} x_2 - \dot{x}_{1d} \\ \dot{x}_4 - \dot{x}_{3d} \\ \dot{x}_6 - \dot{x}_{5d} \end{bmatrix}
$$
(8)

Select the sliding surface as follows:

$$
s = \begin{bmatrix} s_1 \\ s_3 \\ s_5 \end{bmatrix} = \begin{bmatrix} c_1 e_1 + c_2 e_1 \\ c_3 e_3 + c_4 e_3 \\ c_5 e_5 + c_6 e_5 \end{bmatrix} \Rightarrow
$$
  
\n
$$
\dot{s} = \begin{bmatrix} c_1 e_1 + c_2 e_1 \\ c_3 e_3 + c_4 e_3 \\ c_5 e_5 + c_6 e_5 \end{bmatrix} = \begin{bmatrix} c_1 (x_2 - x_{1d}) + c_2 (x_2 - x_{1d}) \\ c_3 (x_4 - x_{3d}) + c_4 (x_4 - x_{3d}) \\ c_5 (x_6 - x_{5d}) + c_6 (x_6 - x_{5d}) \end{bmatrix}
$$
(9)

Substituting  $\dot{x}_2, \dot{x}_4, \dot{x}_6$  into equation (9), we have:

$$
\dot{s} = \begin{bmatrix} c_1(x_2 - x_{1d}) + c_2(h_1x_2 + h_2 + U_x - x_{1d}^*) \\ c_3(x_4 - x_{3d}) + c_4(h_3x_4 + h_4 + U_y - x_{3d}^*) \\ c_5(x_6 - x_{5d}) + c_6(h_5x_6 + h_6 + U_z - x_{5d}^*) \end{bmatrix}
$$
(10)

Select the slide mode controller for the system. In the sliding mode controller, there are two components, which are equilibrium and robust terms, with the formula:

$$
u = u_{EQ} + u_R \tag{11}
$$

The equilibrium component will be selected by setting the sliding surface's derivative  $\dot{s} = -k s$  and giving the system's uncertainty parameter  $= 0$ . Then we have:

$$
\begin{bmatrix} u_{EQx} \\ u_{EQz} \\ u_{EQz} \end{bmatrix} = \frac{\begin{bmatrix} \frac{1}{c_2} (-c_1(x_2 - x_{1d}) - c_2(h_1x_2 - x_{1d}) - k_1s_1) \\ \frac{1}{c_4} (-c_3(x_4 - x_{3d}) - c_4(h_3x_4 - x_{3d}) - k_3s_3) \\ \frac{1}{c_6} (-c_5(x_6 - x_{5d}) + c_6(h_5x_6 - x_{5d}) - k_5s_5) \end{bmatrix}
$$
\n(12)

The robust component will be chosen  $u_R = -\eta sign(s)$ , so we have:

$$
u_R = \begin{bmatrix} u_{Rx} \\ u_{Ry} \\ u_{Rz} \end{bmatrix} = \begin{bmatrix} -\frac{1}{c_2} (\eta_1 sign(s_1)) \\ -\frac{1}{c_4} (\eta_3 sign(s_3)) \\ -\frac{1}{c_6} (\eta_5 sign(s_5)) \end{bmatrix}
$$
(13)

 $[U_2]$ 

 $\eta$  is chosen so that  $\eta \geq$  is the maximum value of the uncertain parameters in the system.

Substituting  $u_{E0}$ ;  $u_R$  into (11) we have:

$$
\begin{bmatrix} U_x \\ U_y \\ U_z \end{bmatrix} = \begin{bmatrix} \frac{1}{c_2}(-c_1(x_2 - x_{1d}) - c_2(h_1x_2 - x_{1d}) - k_1s_1 - \eta_1sign(s_1)) \\ \frac{1}{c_4}(-c_3(x_4 - x_{3d}) - c_4(h_3x_4 - x_{3d}) - k_3s_3 - \eta_3sign(s_3)) \\ \frac{1}{c_6}(-c_5(x_6 - x_{5d}) + c_6(h_5x_6 - x_{5d}) - k_5s_5 - \eta_5sign(s_5)) \end{bmatrix}
$$
\n(14)

After obtaining the control signals  $U_x$ ,  $U_y$ ,  $U_z$  we calculate the control signal  $U_1$  in the system of equations as follows:

$$
U_1 = m\sqrt{U_x^2 + U_y^2 + (U_z + g)^2}
$$
 (15)

# **Prove stability of the controlled system.**

Substituting the sliding mode control rule in expression (11) into (10), we have:

$$
\dot{s} = \begin{bmatrix} \dot{s}_1 \\ \dot{s}_3 \\ \dot{s}_5 \end{bmatrix} = \begin{bmatrix} -k_1 s_1 - \eta_1 \dot{s} \, \dot{s} \, n(s_1) + h_2 \\ -k_3 s_3 - \eta_3 \dot{s} \, \dot{s} \, n(s_3) + h_4 \\ -k_5 s_5 - \eta_5 \dot{s} \, \dot{s} \, n(s_5) + h_6 \end{bmatrix} \tag{16}
$$

To prove the stability of the control system, the Lyapunov function is chosen as follows:

$$
v = \begin{bmatrix} v_1 \\ v_3 \\ v_5 \end{bmatrix} = \begin{bmatrix} \frac{1}{2} s_1^2 \\ \frac{1}{2} s_3^2 \\ \frac{1}{2} s_5^2 \end{bmatrix}
$$
 (17)

Calculating the derivative  $\nu$ , we have:

$$
\begin{bmatrix} \vec{v}_1 \\ \vec{v}_3 \\ \vec{v}_5 \end{bmatrix} = \begin{bmatrix} s_1 \vec{s}_1 \\ s_3 \vec{s}_3 \\ s_5 \vec{s}_5 \end{bmatrix} = \begin{bmatrix} s_1(-k_1 s_1 - \eta_1 \text{sign}(s_1) + h_2) \\ s_3(-k_3 s_3 - \eta_3 \text{sign}(s_3) + h_4) \\ s_5(-k_5 s_5 - \eta_5 \text{sign}(s_5) + h_6) \end{bmatrix}
$$

$$
= \begin{bmatrix} -k_1 s_1^2 - s_1(\eta_1 \text{sign}(s_1) - h_2) \\ -k_3 s_3^2 - s_3(\eta_3 \text{sign}(s_3) - h_4) \\ -k_5 s_5^2 - s_5(\eta_5 \text{sign}(s_5) - h_6) \end{bmatrix}
$$
(18)

We have:

$$
\begin{bmatrix} -k_1 s_1^2 - s_1(\eta_1 \sin(\zeta_1) - h_2) \\ -k_3 s_3^2 - s_3(\eta_3 \sin(\zeta_3) - h_4) \\ -k_5 s_5^2 - s_5(\eta_5 \sin(\zeta_5) - h_6) \end{bmatrix} \le \begin{bmatrix} -k_1 s_1^2 - |s_1|(\eta_1 - h_2) \\ -k_3 s_3^2 - |s_3|(\eta_3 - h_4) \\ -k_5 s_5^2 - |s_5|(\eta_5 - h_6) \end{bmatrix}
$$
  

$$
\le \begin{bmatrix} -k_1 s_1^2 \\ -k_3 s_3^2 \\ -k_5 s_5^2 \end{bmatrix} \le 0
$$

Since we have the condition  $n$ . The  $n$  is chosen so that  $\eta \geq$  is the maximum value of the uncertain parameters in the system. Therefore, according to the Lyapunov stability theory, our system is stable. The calculation process is similar to above, we have:

$$
\begin{bmatrix}\nU_3 \\
U_4\n\end{bmatrix}\n= \n\begin{bmatrix}\n\frac{1}{c_8h_9}[-c_7(x_8 - x_{7d}) - c_8(h_7x_{10}x_{12} + h_8x_{10} - x_{7d}) - k_7s_7 - \eta_7sign(s_7)] \\
\frac{1}{c_{10}h_{13}}[-c_9(x_{10} - x_{9d}) - c_{10}(h_{11}x_8x_{12} + h_{12}x_8 - x_{9d}) - k_9s_9 - \eta_9sign(s_9)] \\
\frac{1}{c_{12}h_{16}}[-c_{11}(x_{12} - x_{11d}) - c_{12}(h_{15}x_8x_{10} - x_{11d}) - k_{11}s_{11} - \eta_{11}sign(s_{11})]\n\end{bmatrix}
$$
\n(19)

#### *B. Reduce Chattering Phenomenon*

Sliding mode control (SMC) has several advantages: robustness against uncertainties and disturbances, high accuracy, finite-time convergence, and simplicity. However, sliding mode control has the disadvantage of the chattering phenomenon because it uses the  $\eta$ sign(s) function in the control signal. Therefore, we need to devise a solution to overcome this phenomenon. In this study, we replace the  $\eta$ sign(s) function with  $\eta |s|^\lambda$ sign(s). The simulation results in Fig. 3 demonstrate the effectiveness of this solution.



Fig. 3. Solution for chattering phenomenon, a) Method without  $\eta |s|^{\lambda}$ sign(s). b) Method with  $\eta |s|^\lambda$ sign(s)

Then formula (14) and formula (19) are rewritten as (20) and (21).

#### *C. Simulation results*

In this section, we have successfully designed a sliding mode controller for position control and attitude control of quadcopter based on Lyapunov stability theory. The chattering phenomenon reduction is also considered. In the next section, this controller will be used for quadcopter trajectory tracking and landing with the parameters shown in Table I.

$$
\begin{bmatrix}\nU_x \\
U_y \\
U_z\n\end{bmatrix} = \begin{bmatrix}\n\frac{1}{c_2}(-c_1(x_2 - x_{1d}) - c_2(h_1x_2 - x_{1d} - h_1s_1 - \eta_1|s_1|^2 \text{sign}(s_1)) \\
\frac{1}{c_4}(-c_3(x_4 - x_{3d}) - c_4(h_3x_4 - x_{3d} - k_3s_3 - \eta_3|s_3|^2 \text{sign}(s_3)) \\
\frac{1}{c_6}(-c_5(x_6 - x_{5d}) + c_6(h_5x_6 - x_{5d} - k_5s_5 - \eta_5|s_5|^2 \text{sign}(s_5))\n\end{bmatrix}
$$
\n(20)  
\n
$$
\begin{bmatrix}\nU_2 \\
U_3 \\
U_4\n\end{bmatrix} = \begin{bmatrix}\n\frac{1}{c_8h_9}[-c_7(x_8 - x_{7d}) - c_8(h_7x_{10}x_{12} + h_8x_{10} - x_{7d}) - k_7s_7 - \eta_7|s_7|^2 \text{sign}(s_7)] \\
\frac{1}{c_{10}h_{13}}[-c_9(x_{10} - x_{9d}) - c_{10}(h_{11}x_8x_{12} + h_{12}x_8 - x_{9d}) - k_9s_9 - \eta_9|s_9|^2 \text{sign}(s_9)] \\
\frac{1}{c_{12}h_{16}}[-c_{11}(x_{12} - x_{11d}) - c_{12}(h_{15}x_8x_{10} - x_{11d}) - k_{11}s_{11} - \eta_{11}|s_{11}|^2 \text{sign}(s_{11})]\n\end{bmatrix}
$$
\n(21)

TABLE I. QUADCOPTER SIMULATION PARAMETERS

Parameters	Value	Unit	Desc		
	9.81	m/s <sup>2</sup>	Gravitational acceleration		
	0.225	m	Distance from quadrotor center to rotor center		
m	$\mathcal{L}$	kg	<b>Ouadrotor</b> mass		
$I_{xx}$	0.0035	kg/m <sup>2</sup>	Moment of inertia of the frame along the x axis.		
$I_{yy}$	0.0035	kg/m <sup>2</sup>	Moment of inertia of the frame along the y axis		
$1_{77}$	0.0050	kg/m <sup>2</sup>	Moment of inertia of the frame along the z axis.		

Fig. 4 and Fig. 5 are the simulation results of the output signals of the quadcopter's state variables. Based on the simulation results, we can see that external disturbances greatly influence the operation of the quadcopter. So, it is necessary to come up with a solution to resist these types of interference and make the quadcopter's operation smoother. Fig. 4 shows the output status signal of the system when the SMC controller is not used. It shows that the flight process is volatile when affected by interference. The simulation results in Fig. 5 show that when using the SMC controller for the above system, the orbital tracking quality of the quadcopter is better, and the stability is greatly improved.



Fig. 4. When external disturbances without using the SMC controller



Fig. 5. When there is an external disturbance, the SMC controller is used

# IV. LANDING CONTROL BASED ON ENERGY **PREDICTION**

We set up the simulation environment of the quadcopter in 3D space (20, 20, 16). Our simulation scenarios include:

Scenario 1: The quadcopter works in normal mode, meaning that its energy is full throughout the operation.

Scenario 2: During operation, bad weather, such as strong winds, requires more energy, causing the quadcopter to no longer have enough to reach the desired destination. It needs to land at a backup charging station to charge until the SoC reaches more than 90% and then continue flying to the desired destination.

The quadcopter flight process has two phases; phase 1 is when the quadcopter follows the trajectory to the desired location according to the GPS signal. Phase 2 is activated when the quadcopter detects the marker. Once the marker is detected, the quadcopter will land based on the error between the current position of the UAV and the marker based on computer vision technology. The quadcopter's landing process depends on energy prediction to avoid risks to the quadcopter. The entire flight and landing process of the quadcopter is carried out according to the algorithm flowchart in Fig. 6.



Fig. 6. Quadcopter flight algorithm flowchart based on energy prediction

Apply the controller built in section 3. Now, we apply to control the quadcopter to follow the trajectory based on the forecast of energy consumption. The SoC value of the battery is updated and monitored in real-time. From there, the algorithm will estimate the distance the quadcopter can travel. If the estimated distance is greater than or equal to the distance to the destination, the quadcopter will operate in mode 1. If the estimated distance is less than the distance to the destination, the quadcopter will activate mode 2 in Fig. 7. The algorithm to find the nearest landing station is activated. When the nearest landing station is found, the quadcopter will create a new trajectory to move there using the SMC controller to follow this new trajectory.

When  $\text{SoC} < \text{SoC}_{threshold}$ , UAV has not reached the target:  $SoC < SoC_{threshold}$  UAV will the emergency landing mode be activated, and then the UAV will calculate and find the nearest landing station. The activation of the nearest station finding algorithm is completely random without any fixed point.



Fig. 7. Marker landing control scheme based on energy prediction

Below are some cases of UAV emergency landing when the energy is no longer enough to fly to the target as originally proposed. We have the positions of 3 markers in space, which are respectively 3 UAV landing stations with coordinates: maker1(10; 0; 0); maker2 (20; 10; 0); marker3 (20; 20; 0). Markers 1 and 2 are backup landing stations (charging stations), and marker 3 is the target landing station the UAV needs to reach during operation as shown in Fig. 8.

Scenario 1: The quadcopter works in normal mode, meaning that its energy is full throughout the operation.

This is a scenario where the UAV operates in normal mode under clear weather conditions, with no unusual power usage. The UAV flies in a pre-set trajectory using the SMC controller in mode 1 in Fig. 7.



Fig. 8. When  $SoC \geq SoC_{threshold}$  UAV reaches the target

Scenario 2: UAVs work in bad weather conditions, such as strong winds and heavy rain, and therefore require more energy. Causing the quadcopter to no longer have enough to reach the desired destination. It needs to land at a backup charging station to charge until the SoC reaches more than 90% and then continue flying to the desired destination.

Distance from UAV→ marker:

$$
S_i^q
$$
  
=  $\sqrt{(x_q - x_{maker}^i)^2 + (y_q - y_{maker}^i)^2 + (z_q - z_{maker}^i)^2}$  (22)

$$
S_{min} = min\{S_i^q\}, i = 1, 2, ...n
$$
 (23)

When the UAV flies to position A (0; 9.584; 15.004), it is predicted that its energy is not enough to fly to the destination. Therefore, it activates the emergency landing algorithm and goes to the nearest charging station to charge more energy until the SoC reaches 90%. Then, the UAV automatically takes off according to the given trajectory to reach the destination.

According to formula (22) and (23), the closest distance from position A to marker 1 is. The simulation results also show that the UAV landed at marker 1, as shown in Fig. 9.

Similarly, in Fig. 10, the UAV flies to position B (10.01; 3.94; 15) and is predicted to have insufficient energy to reach the target point. Therefore, it activates mode 2 in Fig. 7 and finds the closest distance to land at marker station 1.



Fig. 9. When the UAV reaches A (0; 6.169; 15), it is predicted that it is running out of energy and needs to make an emergency landing



Fig. 10. When the UAV reaches B (10.01; 3.94; 15), it is predicted that it is running out of energy and needs to make an emergency landing

Fig. 11 shows that when the UAV flies to position C (10.01; 11.1; 15.0), the UAV is predicted not to have enough energy to reach the destination. Therefore, the mode to find the nearest charging station and make an emergency landing (mode 2 in Fig. 7) is activated at position C. When the battery's SoC reaches 90%, the UAV automatically takes off. It flies to the destination according to the predetermined trajectory.



Fig. 11. When the UAV reaches C (10; 11.1; 15), it is predicted that it is running out of energy and needs to make an emergency landing

Similar to the above cases, the UAV flies to position D (20, 7.7, 15) and activates mode 2 in the diagram in Fig. 7. The algorithm finds marker 2 with the closest distance to the UAV, so the UAV lands at marker 2, as shown in Fig. 12.



Fig. 12. When the UAV reaches D (20; 7.7; 15), it is predicted that it is running out of energy and needs to make an emergency landing

#### V. EXPEERIMENTAL RESULTS

In this study, we have only conducted experiments with the quadcopter's trajectory-tracking ability when using SMC and PID controllers. To experimentally verify the control algorithm, we tested the F450 quadcopter's ability to fly according to a given trajectory automatically in a real environment. The parameters of quadcopter F450 are given in Table II.

We simulate a space with dimensions of 20m wide, 20m long, and 20m high. In Fig. 14 and Fig. 15, the blue line is the desired trajectory and is loaded into the quadcopter's memory.

TABLE II. QUADCOPTER SIMULATION PARAMETERS

Parameters	Values		
Mass	$2.3 \text{ kg}$		
Maximum current per motor	$I_{motor} = 4.28 A$		
Current of other loads on the UAV	$I_{other} = 6.2 A$		
Serial cell number in PIN	$N_{series} = 3s$ Cell		
Nominal voltage of the PIN	$V_{bat\,nominal} = 11.1 V$		
Battery capacity	$Q = 4200$ mmAh		
<b>Battery C-rate</b>	$C_{\text{rate}} = 10$		
Battery discharge rule	$DR = 80%$		
Flying load	$L_{flying} = 30\%$		



Fig. 13. Quadcopter F450

The orange line is the flight trajectory performed using the PID algorithm. The green line is the flight trajectory performed using the proposed SMC algorithm.



Fig. 14. Experimental results based on GPS data

All were carried out under the same environmental conditions of no rain and light wind. Looking at the trajectory data returned from the actual flight process in Fig. 14, we can see that using the PID algorithm gives much poorer trajectory control quality than the SMC algorithm we propose.

Table III lists the maximum trajectory tracking error in the x, y, and z directions of the quadcopter during the test corresponding to stages E1 to E7. From this table, we can see that the maximum trajectory tracking error in general in all directions of the SMC is always less than or equal to that of the PID controller.

The test scenario with the flight trajectory is as follows: from the starting point, the quadcopter flies vertically to a height of 15m at a speed of 2.05m/s, then maintains the same altitude and x direction, flies forward in the y direction for 20m at a speed of 1.01m/s; then maintains the same altitude and y direction, flies forward in the x direction for 10m at a speed of 1.01m/s; then maintains the same altitude and x direction, flies back in the y direction for 20m at a speed of 1.21m/s; then maintains the same altitude and y direction, flies forward in the x direction for 10m at a speed of 0.73m/s; then maintains the same altitude and x direction, flies forward in the y direction for 20m at a speed of 1.21m/s; Finally keep x and y directions constant and land at 0.59m/s.

TABLE III. ERRORS IN DIFFERENT DIRECTION DURING THE QUADCOPTER TEST FLIGHT

Stage	$Error_x(m)$		$Error_y(m)$		$Error_z(m)$				
	PID	<b>SMC</b>	<b>PID</b>	<b>SMC</b>	<b>PID</b>	<b>SMC</b>			
E1(z)	0.15	0.15	1.75	0.5	2	1.5			
E2(y)	0.75	0.4	1.5	1.0	0.1	0.1			
E3(x)	0.75	0.1	0.75	0.75	0.1	0.1			
E4(y)	0.5	0.5	1.0	0.75	0.1	0.1			
E5(x)	1.75	0.75	2.0	1.4	0.1	0.1			
E6(y)	0.5	0.25	2.0	1.6	0.1	0.1			
E7(z)	0.75	0.5	1.0	0.5	$\mathcal{F}$	0.6			
E1: At the starting point, the quadcopter, keeping the x and y directions the same, flies vertically in the z direction to a height of 15m at a speed of $1.55m/s$ . E2: Keeping the z and x directions the same, flies forward in the y direction 20m at a speed of $1.01 \text{m/s}$ ; E3: Keeping the z and y directions the same, flies forward in the x direction 10m at a speed of 1.01m/s. E4: Keeping the z and x directions the same, flies backward in the y direction 20m at a speed of 1.24m/s. E5: Keeping the z and y directions the same, flies forward in the x direction 10m at a speed of 0.73m/s. E6: Keeping the z and x directions the same, flies forward in the y direction 20m at a speed of $1.21 \text{m/s}$ . E7: Keeping x and y directions constant, land in the z-direction to the parking point at a speed of 0.59m/s.									



Fig. 15. The position of the quadcopter in the x direction corresponds to three cases: desired trajectory, PID-controlled trajectory, and sliding mode control trajectory

More specifically, as shown in Fig. 15 and Table III, the trajectory tracking error in the x direction in stage E1 has a relatively small and similar value, about 0.15m, for both PID and sliding controllers. Next, stage E2's largest trajectory tracking error in the x direction is 0.4m for the sliding control and 0.75m for the PID control. In stage E3, the trajectory tracking error of the PID controller tends to increase to 0.75m at the end of the stage, while the sliding controller gives a trajectory tracking error of 0.1m. In stage E4, the largest trajectory tracking error in the x direction of the sliding control and PID is almost equal to 0.5m. In stage E5, the largest trajectory tracking error in the x direction is 1.75m for the PID and 0.75m for the sliding controller. The E6 phase has the largest error of 0.5m for the PID controller and 0.25m for the glide control. The landing phase, E7, has the largest trajectory tracking error of the glide control of 0.5m and 0.75m for the glide and PID controllers, respectively.

Fig. 16 shows that the trajectory tracking stability control quality of the sliding mode controller is better than that of the PID control, especially during the stationary stability process at the y-direction position.

In Fig. 16 and the data in Table III, the E1 stage has the largest y-direction error for the PID controller at 1.75m, while the SMC remains stable in the y-direction with a maximum error of 0.5m. Next, in the E2 stage, the quadcopter moves forward in the y-direction for 20m; at this time, the largest ydirection tracking error of the PID controller is 1.5m and 1m for the SMC. In the E3 stage, the largest y-direction tracking error is 0.75m for both controllers. In the E4 stage, the largest y-direction tracking error is 1m for the PID controller and 0.75m for the SMC. In the next E5 stage, the PID controller has the largest tracking error of 2m, while the SMC controller has 1.4m. In stage E6, this error is 2m for the PID controller and 1.6m for the glide controller. In stage E7, the landing stage, the maximum y-direction tracking error of the PID is 1m, while for the SMC, this error is 0.5m.

The position of the quadcopter in the z direction and the position error in the z direction are shown in Fig. 17 and Table III. The z-direction is kept stable at the height of 15m in stages E2, E3, E4, E5, E6, and changes in stages E1 and E7. During the altitude holding stages, the maximum errors of the PID controller and the SMC are similar and equal to 0.1m. However, during the take-off stage E1, the maximum position tracking error in the z direction of the PID controller

is approximately 2m and is 1.5m for the SMC. During the landing stage E7, these errors are 3m and 0.6m for the PID controller and the SMC, respectively.



Fig. 16. The position of the quadcopter in the y direction corresponds to three cases: desired trajectory, PID-controlled trajectory, and sliding mode control trajectory



Fig. 17. The position of the quadcopter in the z direction corresponds to three cases: desired trajectory, PID-controlled trajectory, and sliding mode control trajectory

# VI. CONCLUSIONS

In this paper, the sliding mode control (SMC) controller is used to control the position and attitude of the quadcopter. Firstly, the dynamic model of the quadcopter is introduced and fully constructed under the consideration of external disturbances. Then, the controller is proposed and developed. The SMC controller is built based on the Lyapunov function so that the system's stability is mathematically guaranteed. The flight process of the quadcopter is carried out in two modes based on the real-time prediction of the quadcopter's energy consumption to avoid risks and support the quadcopter to complete the task under various conditions. The simulation results show the controller can achieve trajectory-tracking feedback with good accuracy and response time.

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